

# Dielectric Constant and Thickness Measurement of Dry Snowpack and Lake Icepack using Correlation Radiometry



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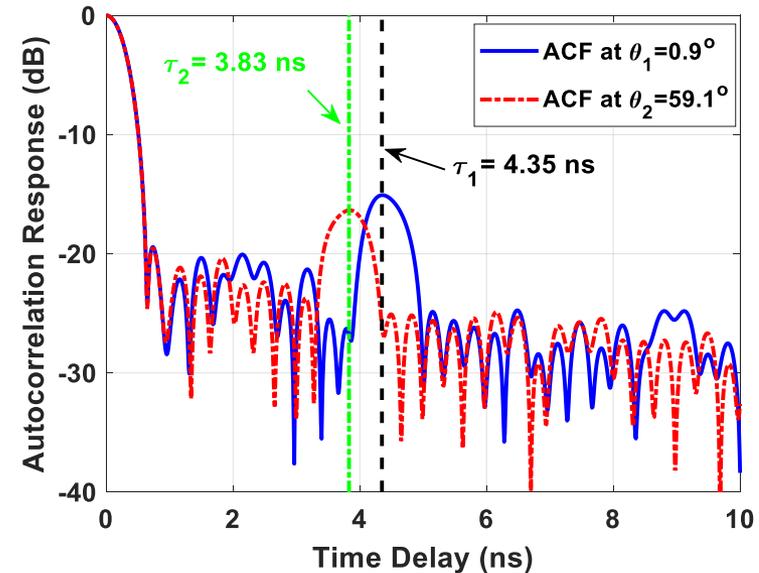
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- Using the time delays measured by WiBAR at two distinct incidence angles:

$$\epsilon_{ice} = \frac{\tau_1^2 \sin^2 \theta_2 - \tau_2^2 \sin^2 \theta_1}{\tau_1^2 - \tau_2^2} = 3.24$$

$$d_{ice} = \frac{c\tau_1}{2\sqrt{\epsilon_p - \sin^2 \theta_1}} = 36.24 \text{ cm}$$



- The measured refractive index and thickness of the freshwater icepack by WiBAR was 3.24 and 36.24 cm ( $\frac{\delta d}{d} \approx 2\%$ ,  $\frac{\delta \epsilon_r}{\epsilon_r} \approx 1.8\%$ ). The ground truth was about 35.56 cm. The dielectric constant of a freshwater icepack  $\epsilon_{ice} = 3.18$  over microwave frequencies (Matzler and Wegmuller<sup>1</sup>).

<sup>1</sup>C. Matzler and U. Wegmuller, "The dielectric properties of fresh-water ice at microwave frequencies," Journal of Physics, D: Applied Physics, vol. 20, pp. 1623-1630, 14 Dec. 1987.

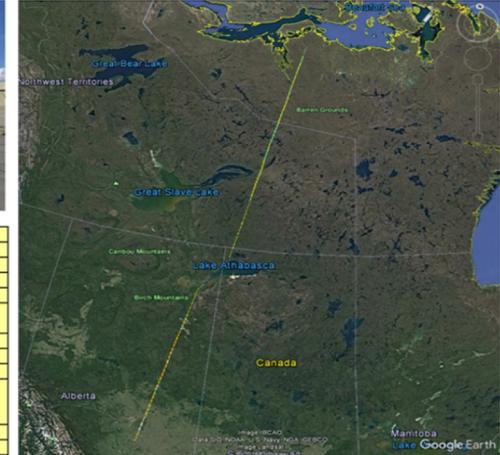
# Future Plans

## Sky is the limit!

- Ultra-Wideband Microwave Radiometer (UWBRAD)
  - The Ohio State University, Professor Joel Johnson's Group



Freq (GHz)	0.5-2, 12 x ~ 88 MHz channels
Polarization	Single (Right-hand circular)
Observation angle	Nadir
Spatial Resolution	1.2 km x 1.2 km (1 km platform altitude)
Integration time	100 msec
Ant. Gain (dB) / Beamwidth	11 dB / 60°
Calibration (Internal)	Reference load and Noise diode sources
Calibration (External)	Sky and Ocean Measurements
Noise equiv dT	0.4 K in 100 msec (each 88 MHz channel)
Interference Management	Full sampling of 88 MHz bandwidth in 16 bits resolution each channel; real time "software defined" RFI detection and mitigation
Initial Data Rate	700 Megabytes per second (10% duty cycle)
Data Rate to Disk	<1 Megabyte per second



- Finally from space!





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# Error Analysis

- The lower bound on the variance of an unbiased estimator can be given by Cramer-Rao lower bound (CRLB):

$$\text{Var}_{\tau_{\text{delay}}}(N) \geq \frac{3\sigma_w^2}{(2\pi A_e \bar{e})^2 N (N - 1)(2N - 1)}$$

where  $\sigma_w^2$  is the variance of the added white Gaussian noise. If an estimator reaches this lower bound, it will be the minimum-variance unbiased (MVU) estimator.

- Since the time delay estimation using IFFT is a nonlinear problem, it will only reach the CRLB for signal to noise ratios (SNR) higher than a threshold value.

- IFFT:

$$S[k] = \frac{1}{N} \sum_{n=0}^{N-1} e[n] \exp\left(i \frac{2\pi}{N} kn\right) \quad e[n] = \bar{e}(1 - 2A_e \cos(2\pi f_n \tau_{delay}))$$

- Denote the ACF coefficients at  $\tau = \frac{k}{N}$  for the noise only and the signal present cases by  $S_{e_0}(k)$  and  $S_{e_1}(k)$ , respectively. Hence:

$$S_{e_1}(k_0) = \sqrt{A^2 + B^2}$$

$$A \triangleq \frac{1}{N} \sum_{n=0}^{N-1} (\bar{e} - 2A_e \bar{e} \cos(2\pi n \tau_{delay}) + w[n]) \cos\left(\frac{2\pi n k_0}{N}\right)$$

$$B \triangleq \frac{1}{N} \sum_{n=0}^{N-1} (\bar{e} - 2A_e \bar{e} \cos(2\pi n \tau_{delay}) + w[n]) \sin\left(\frac{2\pi n k_0}{N}\right)$$

where  $A$  and  $B$  are independent Gaussian random variables and of identical variances of  $\sigma^2 = \frac{\sigma_w^2}{2}$ , where  $\sigma_w^2$  is the power of the white Gaussian noise  $w[n]$ .

- The PDF of  $S_{e_1}(k_0)$ , which is Ricean, is given by:

$$\rho_s(u) = \frac{u}{\sigma^2} \exp\left(-\frac{s^2 + u^2}{2\sigma^2}\right) I_0\left(\frac{u s}{\sigma^2}\right)$$

Where  $s = A_e \bar{e}$  is the square root of the sum of the mean square of  $A$  and  $B$ , and  $I_0$  is the modified Bessel function of the first kind of order zero.

- Similarly, the PDF of the remaining  $S_{e_1}(k)$  and all the  $S_{e_0}(k)$  are all of Rayleigh distribution:

$$\rho_w(u) = \frac{1}{\sigma^2} \exp\left(-\frac{u^2}{2\sigma^2}\right)$$

- The mean square time delay error (MTDE) of the IFFT approach is given by:

$$MSTDE = (1 - q)Var_{\tau_{delay}}(N) + q \int_0^{0.5} 2(u - \tau_{delay})^2 du$$

where  $q$  is the probability of occurrence of an anomaly in the IFFT, as given by:

$$\begin{aligned} q &= P\left(S_{e_1}(k_0) \leq \text{at least one of the } S_{e_1}(1), S_{e_1}(2), \dots, S_{e_1}(k_0 - 1), S_{e_1}(k_0 + 1), \dots, S_{e_1}\left(\frac{N}{2} - 1\right)\right) \\ &= \int_0^\infty P(S_{e_1}(k_0)) \cdot \left[ 1 - \prod_{k=1, k \neq k_0}^{\left(\frac{N}{2}\right)-1} P(S_{e_1}(k) < u) \right] du \\ &= \int_0^\infty \rho_S(u) \left[ 1 - \prod_{k=1, k \neq k_0}^{\left(\frac{N}{2}\right)-1} \int_1^2 \rho_W(v) dv \right] du \end{aligned}$$

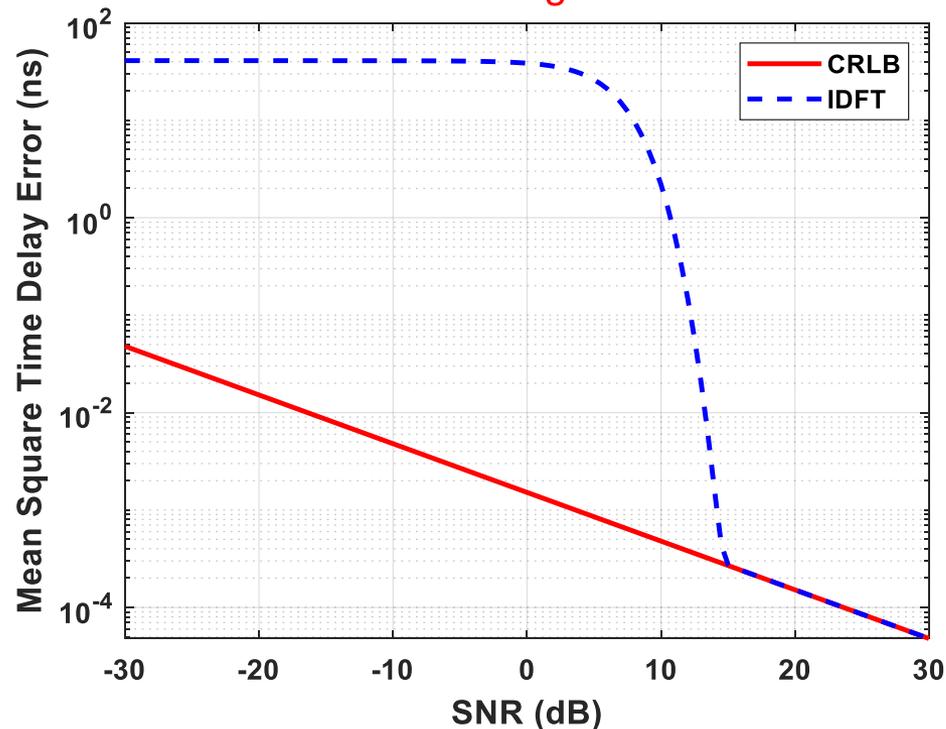
# Non-Destructive Dielectric Constant Measurement of a Loss-Less Dielectric Slab using FD-WiBAR

- If we define  $SNR = \frac{\left(\frac{Aee}{2}\right)^2}{\sigma_w^2}$ , the CRLB would be:

$$Var_{\tau_{delay}}(N) \geq \frac{3}{16\pi^2 N (N - 1)(2N - 1)SNR}$$

- Theoretical time delay variance of IFFT for an icepack with  $\tau_{delay} = 3.83$  ns with respect to the SNR of the delay peak:

The incidence angle is  $\theta = 0^\circ$ .



The bandwidth is 3 GHz,  
and the number of points  
is  $N = 461$ .

- Measured time delay by WiBAR at two distinct incidence angle:

$$\begin{cases} \tau_1 = \tau_{delay}(\theta_1) = \frac{2d_p}{c} \sqrt{\epsilon_p - \sin^2 \theta_1} \\ \tau_2 = \tau_{delay}(\theta_2) = \frac{2d_p}{c} \sqrt{\epsilon_p - \sin^2 \theta_2} \end{cases} \Rightarrow d_p \text{ and } \epsilon_p \text{ can be found}$$

- Assuming  $\epsilon_p$  is real (Loss-Less pack):

$$\epsilon_p = \frac{\tau_1^2 \sin^2 \theta_2 - \tau_2^2 \sin^2 \theta_1}{\tau_1^2 - \tau_2^2}$$

- Using measured  $\epsilon_p$ :

$$d_p = \frac{c\tau_i}{2\sqrt{\epsilon_p - \sin^2 \theta_i}}, \quad i = 1, 2$$

- **Objective:**

- To remotely measure the snowpack depth, density, and thus snow water equivalent (SWE)

- **Importance of Snow and Ice:**

- Snow is life!

- **Necessity of monitoring snow and ice:**

- Significant changes in snow accumulation, timing, and melt

- **Method:**

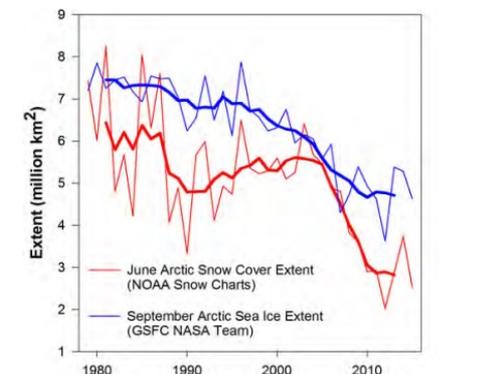
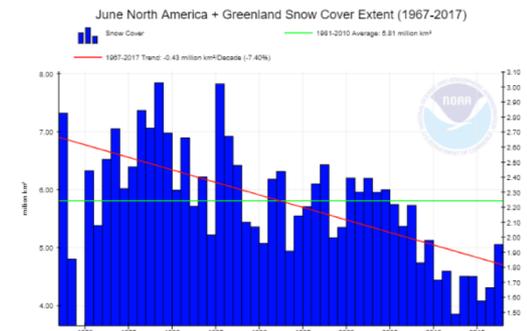
- Remotely sense the propagation time  $\tau_{delay}$  of multi-path microwave emission of low-loss terrain covers at two distinct incident angles using a wideband autocorrelation radiometry (WiBAR)

- **Benefits:**

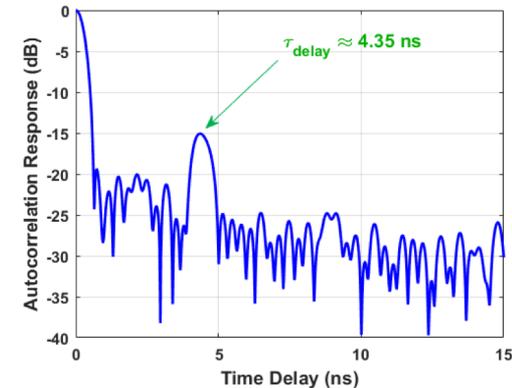
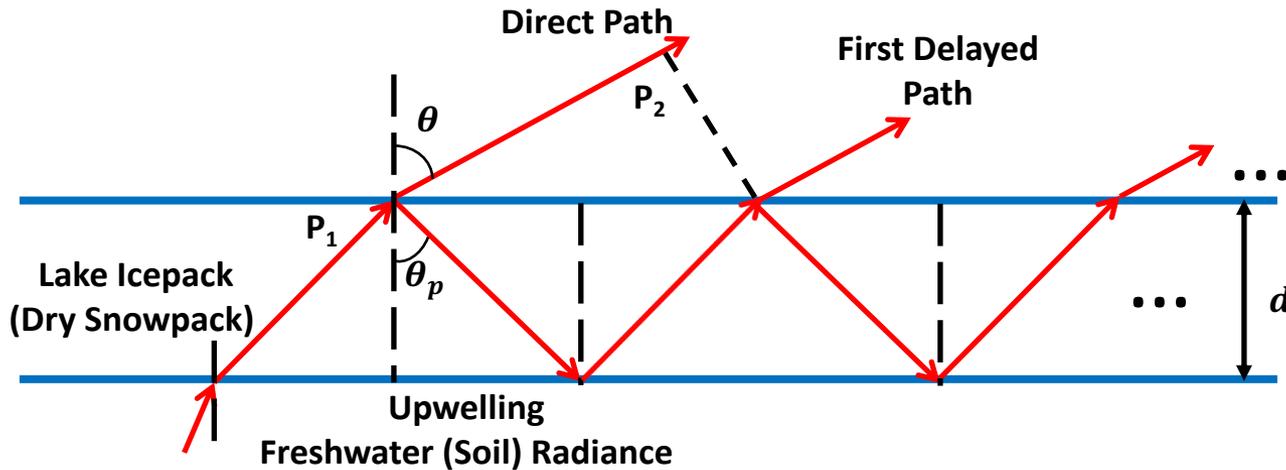
- Passive microwave system
  - ✓ Low power (low cost) for space/air borne instruments
  - ✓ All weather operation capability
- Deterministically measures the thickness and dielectric constant
  - ✓ No algorithm calibration needed

- **Challenges:**

- Requires wide frequency bandwidth
- Radio Frequency Interference (RFI) could be problematic



Sea ice and snow cover extents for the Northern Hemisphere from 1979 to 2015. The thick lines are 5-year running means (Derksen and Brown 2012). 12



This method (WiBAR) uses the coherence of the microwave emission from beneath the pack.

Integrated time delay:

$$\tau_{delay} = \frac{2}{c} \int_0^d \sqrt{\epsilon_p(z) - \sin^2 \theta} dz$$

$\epsilon_p$  is the dielectric constant of the low-loss pack.